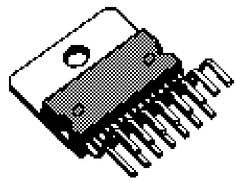
SGS-THOMSON MICROELECTRONICS

L295

DUAL SWITCH-MODE SOLENOID DRIVER

PRELIMINARY DATA

- HIGH CURRENT CAPABILITY (up to 2.5A per channel)
- HIGH VOLTAGE OPERATION (up to 46V for power stage)
- HIGH EFFICIENCY SWITCHMODE OPERATION
- REGULATED OUTPUT CURRENT (adjustable)
- FEW EXTERNAL COMPONENTS
- SEPARATE LOGIC SUPPLY
- THERMAL PROTECTION



Multiwatt 15

ORDER CODE : L295

DESCRIPTION

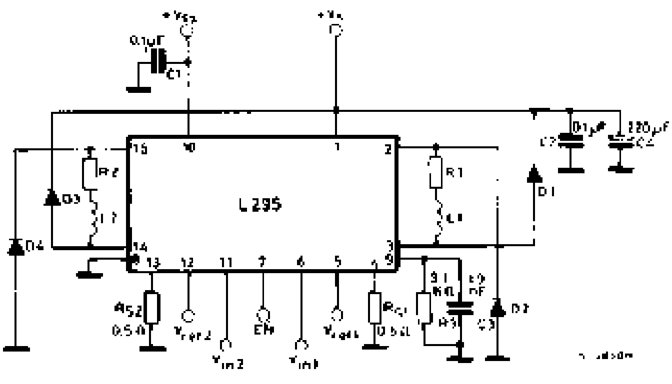
The L295 is a monolithic integrated circuit in a 15 - lead Multiwatt ® package; it incorporates all the functions for direct interfacing between digital circuitry and inductive loads. The L295 is designed to accept standard microprocessor logic levels at the inputs and can drive 2 solenoids. The output current is completely controlled by means of a switching technique allowing very efficient operation. Furthermore, it includes an enable input and dual supplies (for interfacing with peripherals running at a higher voltage than the logic).

The L295 is particularly suitable for applications such as hammer driving in matrix printers, step motor driving and electromagnet controllers.

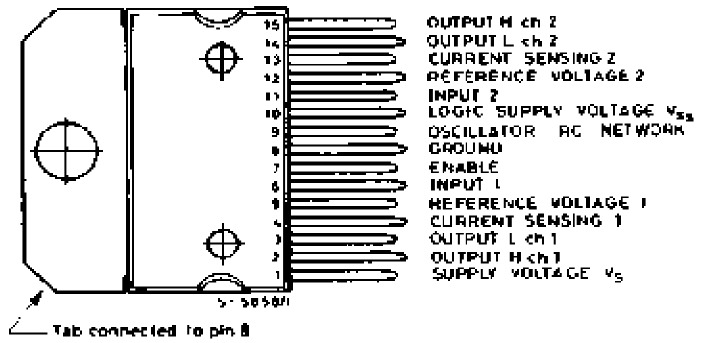
ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V_s	Supply voltage	50	V
V_{ss}	Logic supply voltage	12	V
V_{EN}, V_i	Enable and input voltage	7	V
V_{ref}	Reference voltage	7	V
Io	Peak output current (each channel)		
	- non repetitive (t = 100 µsec)	3	A
	- repetitive (80% on - 20% off; T_{on} = 10 ms)	2.5	A
	- DC operation	2	A
Ptot	Total power dissipation (at Tcase = 75 °C	25	W
Tstg, Tj	Storage and junction temperature	- 40 to 150	°C

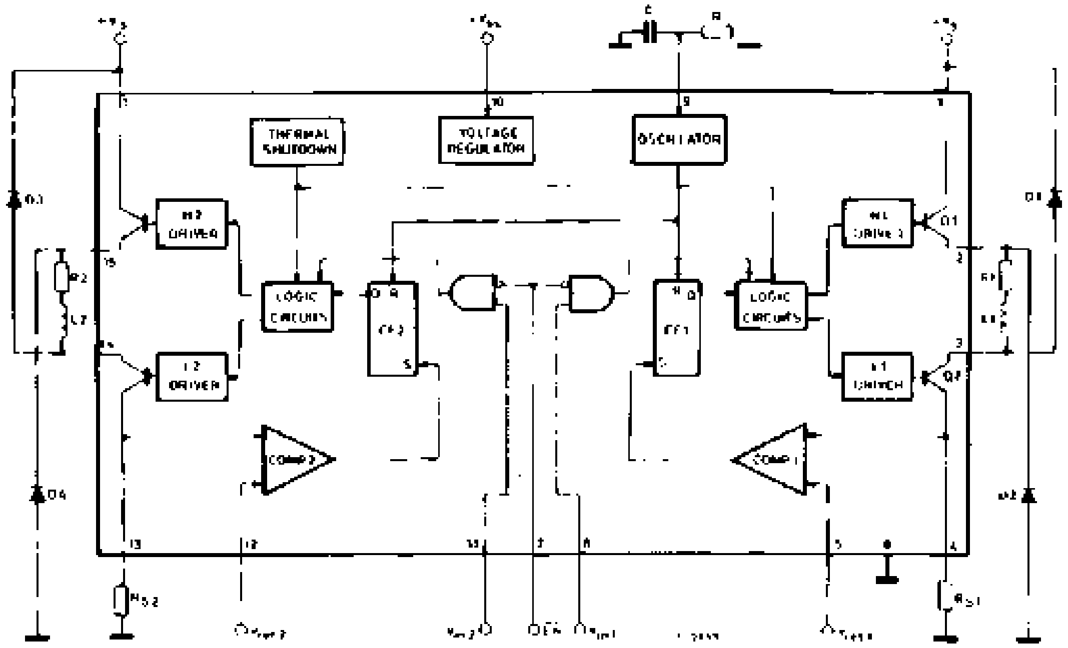
APPLICATION CIRCUIT



CONNECTION DIAGRAM (top view)



BLOCK DIAGRAM



THERMAL DATA

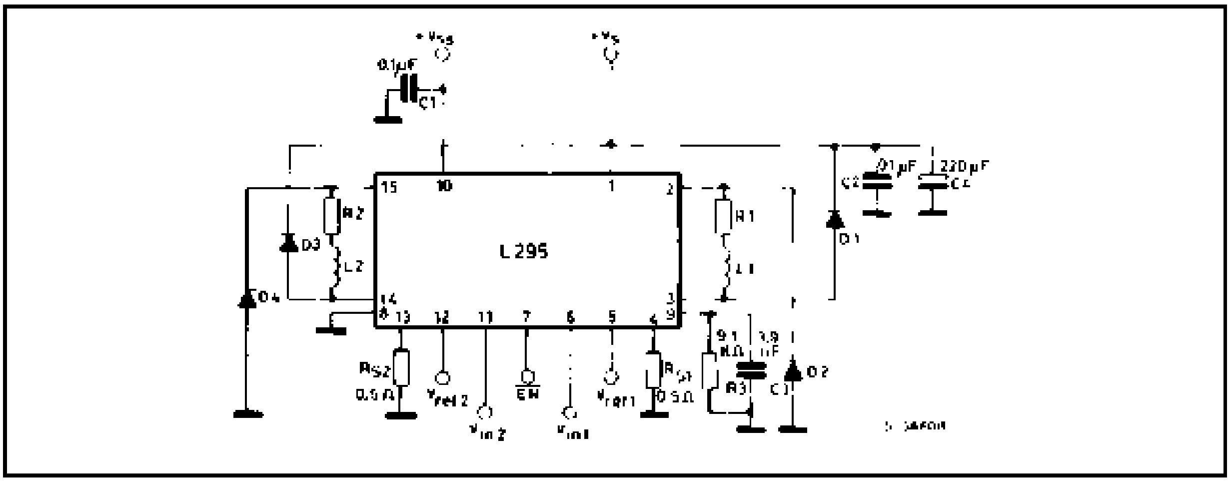
Symbol	Parameter		Value	Unit
$R_{th\text{-}j\text{-}case}$	Thermal resistance junction-case	max	3	°C/W
$R_{th\text{-}j\text{-}amb}$	Thermal resistance junction-ambient	max	35	°C/W

ELECTRICAL CHARACTERISTICS (Refer to the application circuit, V_{ss} = 5V, V_s = 36V; T_j = 25°C; L = Low; H = High; unless otherwise specified)

Symbol	Parameter	Test conditions		Min.	Typ.	Max.	Unit
V_s	Supply Voltage			12		46	V
V_{ss}	Logic Supply Voltage			4.75		10	V
I_d	Quiescent drain current (from VSS)	V_S = 46V; V_{i1} = V_{i2} = V_{EN} = L				4	mA
I_{ss}	Quiescent drain current (from VS)	V_{SS} = 10 V				46	mA
V_{i1},,V_{i2}	Input Voltage		Low	-0.3		0.8	V
			High	2.2		7	
V_{EN}	Enable Input Voltage		Low	-0.3		0.8	V
			High	2.2		7	
I_{i1}, I_{i2}	Input Current		V_{i1} = V_{i2} = L			-100	µA
			V_{i1} = V_{i2} = H			10	
I_{EN}	Enable Input Current		V_{EN} = L			-100	µA
			V_{EN} = H			10	
V_{ref1}, V_{ref2}	Input Reference Voltage			0.2		2	V
I_{ref1}, I_{ref2}m	Input Reference Voltage					-5	µA
F_{osc}	Oscillation Frequency	C = 3.9 nF;	R = 9.1 KΩ		25		KHz
$\frac{I_p}{V_{ref}}$	Transconductance (each ch.)	V_{ref} = 1V		1.9	2	2.1	A/V
V_{drop}	Total output voltage drop (each channel) (*)	I_o = 2 A			2.8	3.6	V
V_{sens1} V_{sens2}	External sensing resistors voltage drop					2	V

(*) V_{drop} = $V_{CEsat\ Q1}$ + $V_{CEsat\ Q2}$.

APPLICATION CIRCUIT



D2, D4 = 2A High speed diodes
D1, D3 = 1A High speed diodes) trr ≤ 200 ns

R1 = R2 = 2Ω
L1 = L2 = 5 mH

FUNCTIONAL DESCRIPTION

The L295 incorporates two indipendent driver channals with separate inputs and outputs, each capable of driving an inductive load (see block diagram).

The device is controlled by three micriprocessor compatible digital inputs and two analog inputs.

These inputs are:

$\overline{EN}$ chip enable (digital input, active low), enables both channels when in the low state.

V_{in1}, V_{in2} channel inputs (digital inputs, active high), enable each channel independently. A channel is actived when both $\overline{EN}$ and the appropriate channel input are active.

V_{ref1}, V_{ref2} reference voltages (analog inputs), used to program the peak load currents. Peak load current is proportional to V_{ref} .

Since the two channels are identical, only channel one will be described.

The following description applies also the channel two, replacing FF2 for FF1, V_{ref} for V_{ref1} etc.

When the channel is avtivated by low level on the EN input and a high level on the channel input, V_{in2}, the output transistors Q1 and Q2 switch on and current flows in the load according to the exponential law:

$$I = \frac{V}{R1} \left(1 - e^{\frac{-R1\ t}{L1}} \right)$$

where: R1 and R2 are the resistance and inductance of the load and V is the voltage available on the load (V_s - V_{drop} - V_{sense}).

The current increases until the voltage on the external sensing resistor, R_{S1}, reaches the reference voltage, V_{ref1}. This peak current, I_{p1}, is given by:

$$I_{p1} = \frac{V_{ref1}}{R_{S1}}$$

At this point the comparator output, Vomp1, sete the RS flip-flop, FF1, that turns off the output transistor, Q1. The load current flowing through D2, Q2, R_{S1}, decreases according to the law:

$$I = \left(\frac{V_A}{R_1} + I_{p1} \right) e^{\frac{-R1\ t}{L1}} - \frac{V_A}{R1}$$

where $V_A = V_{CEsat\ Q2} + V_{sense} + V_{D2}$

If the oscillator pin (9) is connected to ground the load current falls to zero as shown in fig. 1.

At this time t_2 the channel 1 is disabled, by taking the inputs V_{in1} low and/or $\overline{EN}$ high, and the output transistor Q2 is turned off. The load current flows through D2 and D1 according to the law:

$$I = \left(\frac{V_B}{R_1} + I_{T2} \right) e^{\frac{-R1\ t}{L1}} - \frac{V_B}{R1}$$

where $V_B = V_S + V_{D1} + V_{D2}$

I_{T2} = current value at the time t_2.

Fig. 2 in shows the current waveform obtained with an RC network connected between pin 9 and ground. From to t_1 the current increases as in fig. 1. A difference exists at the time t_2 because the current starts to increase again. At this time a pulse is produced by the oscillator circuit that resets the flip.flop, FF1, and switches on the outout transistor, Q1. The current increases until the drop on the sensing resistor R_{S1} is equal to V_{ref1} (t_3) and the cycle repeats.

The switching frequency depends on the value R and C, as shown in fig. 4 and must be chosen in the range 10 to 30 KHz.

It is possible with external hardware to change the reference voltage V_{ref} in order to obtain a high peak current I_p and a lower holding current I_h (see fig. 3).

The L295 is provided with a thermal protection that switches off all the output transistors when the junction temperature exceeds 150°C. The presence of a hysteresis circuit makes the IC work again aftera fall of the junction temperature of about 20°C.

The analog input pins (V_{ref1} , V_{ref2}) can be left open or connected to V_{ss}; in this case the circuit works with an internal reference voltage of about 2.5V and the peak current in the load is fixed only by the value of R_S:

$$I_p = \frac{2.5}{R_S}$$

SIGNAL WAVEFORMS

Figure 1. Load current waveform with pin 9 connected to GND.

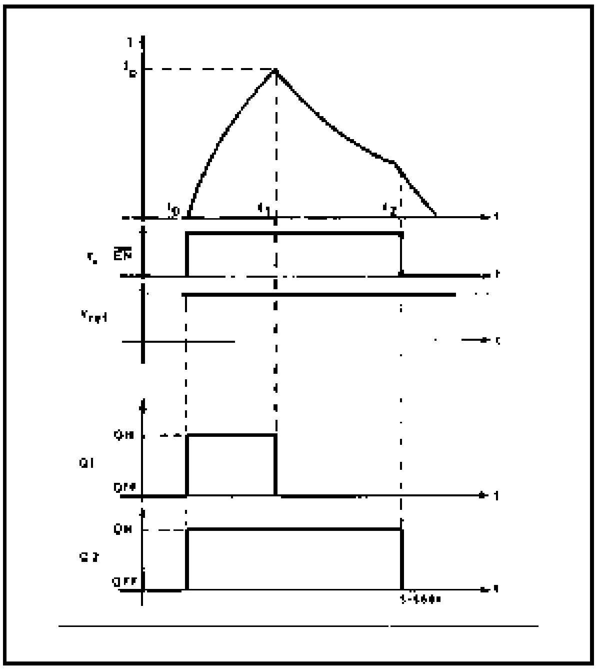
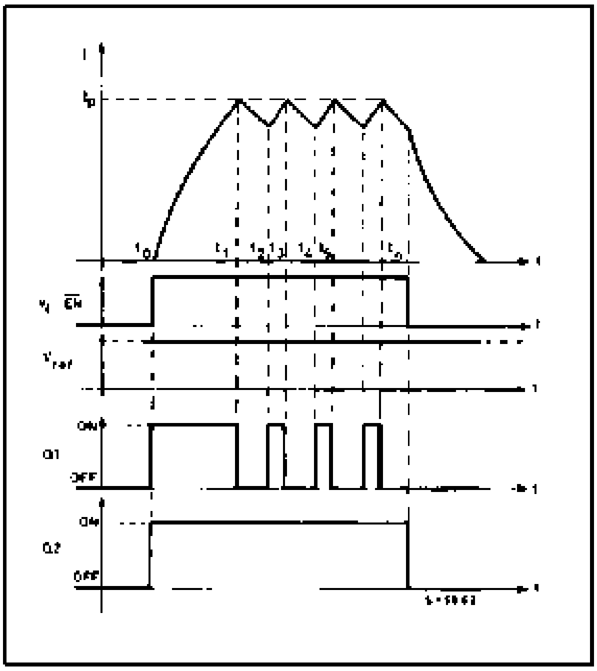


Figure 2. Load current waveform with external R-C network connected between pin 9 and ground.



SIGNAL WAVEFORMS (continued)

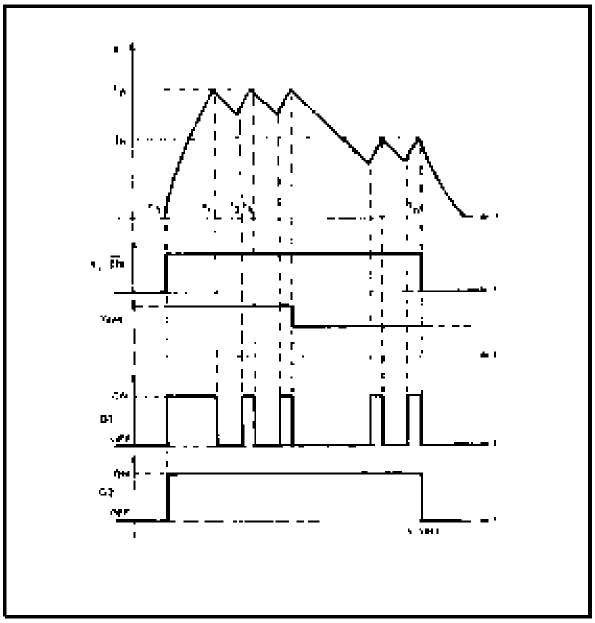


Figure 3. With V_{ref} changed by hardware.

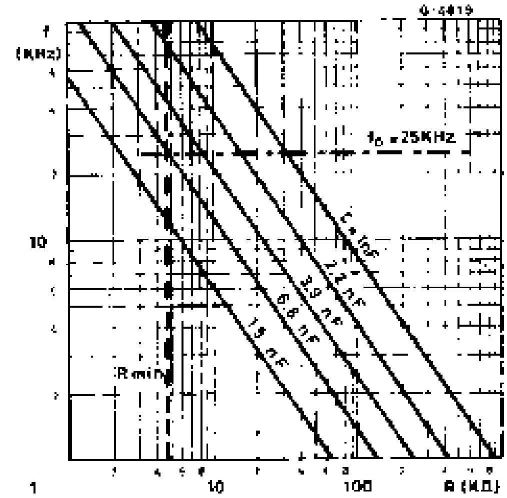
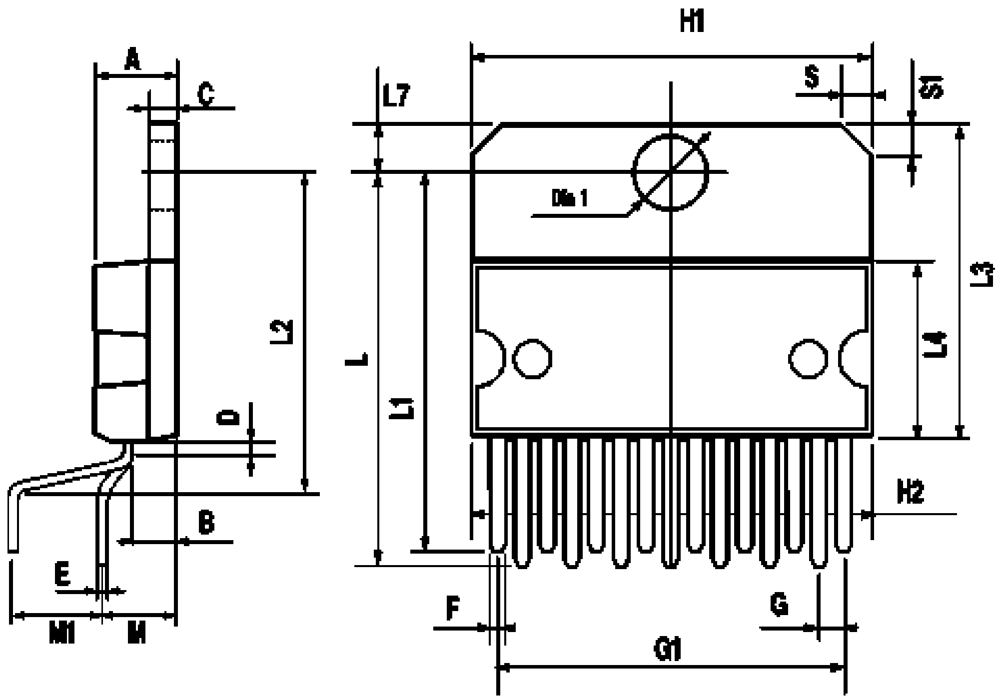


Figure 4. Switching frequency vs. values of R and C.

MULTIWATT15 PACKAGE MECHANICAL DATA

DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A			5			0.197
B			2.65			0.104
C			1.6			0.063
D		1			0.039	
E	0.49		0.55	0.019		0.022
F	0.66		0.75	0.026		0.030
G	1.02	1.27	1.52	0.040	0.050	0.060
G1	17.53	17.78	18.03	0.690	0.700	0.710
H1	19.6			0.772		
H2			20.2			0.795
L	21.9	22.2	22.5	0.862	0.874	0.886
L1	21.7	22.1	22.5	0.854	0.870	0.886
L2	17.65		18.1	0.695		0.713
L3	17.25	17.5	17.75	0.679	0.689	0.699
L4	10.3	10.7	10.9	0.406	0.421	0.429
L7	2.65		2.9	0.104		0.114
M	4.25	4.55	4.85	0.167	0.179	0.191
M1	4.63	5.08	5.53	0.182	0.200	0.218
S	1.9		2.6	0.075		0.102
S1	1.9		2.6	0.075		0.102
Dia1	3.65		3.85	0.144		0.152



SGS-THOMSON Microelectronics GROUP OF COMPANIES

Australia - Brazil - France - Germany - Hong Kong - Italy - Japan - Korea - Malaysia - Malta - Morocco - The Netherlands - Singapore - Spain - Sweden - Switzerland - Taiwan - Thaliand - United Kingdom - U.S.A.